

## 2.6W Mono Class-D Audio Amplifier

### Features

- Supply Voltage Range: 2.5~5.5V
- High Efficiency up to 90%
- Low Power consumption
  - Shutdown Current: 1uA
  - Quiescent Current: 3mA
- High Audio Performance
  - Maximum THD+N: 0.01%
  - Maximum Output Power: 2.6W
- Low-EMI: Unique PWM Control Method
- Resistive Gain Control : 6dB~24dB
- Short & Thermal Protection
- "Pop and Click" Noise Suppression
- Pb-Free Packages are Available

### Descriptions

The CPA011M is a high efficiency filter-free class-D audio power amplifier of delivering 2.6W of continuous average power to a 4Ω from a 5.5V supply in a Bridge Tied Load (BTL) configuration. Under same conditions, the amplifier can provide 1.4W to an 8Ω BTL load with less than 1% THD+N. For portable applications it offers space and cost savings because no output filter is required when using inductive speakers. With more than 90% efficiency and very low shutdown current, it increases the life time of your battery.

The CPA011M processes analog inputs with the unique pulse-width modulation method that lowers output noise and EMI. The gain can be reduced by external input resistors. The CPA011M provides thermal and short circuit protection.

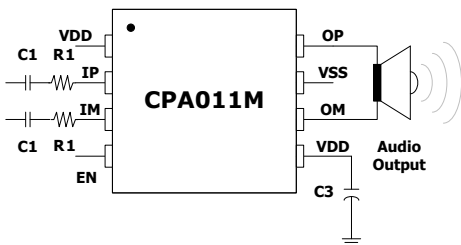
### Applications

- Cellular Phones
- MP3 Players
- PDAs and Smart Phones
- Portable Audio

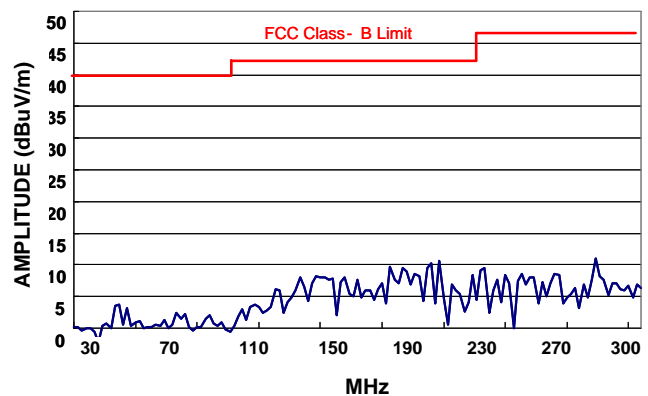
### Package Info.

- 8 DFN 2x2 mm available
- ( The smallest molded package in the world )**

### Application Circuit



### EMI Spread Spectrum



## Absolute Maximum Ratings

Over operating free-air temperature range unless otherwise noted

Symbol	Rating	Max	Unit
VDD	Supply Voltage	6	V
IP/IM	Input Voltage	-0.3 to VDD+0.3	V
T <sub>A</sub>	Operating free-air temperature range	-40 to 85	°C
T <sub>J</sub>	Operating junction temperature range	-40 to 150	°C
Tstg	Storage temperature range	-65 to 85	°C
	ESD Protection		
	Human Body Model (HBM) (Note 1)	>2000	V
	Machine Model (MM) (Note2)	>200	
	Charged Device Model(CDM)	>500	

Stresses exceeding those listed under absolute maximum ratings may cause permanent damage to device.

1. Human Body Model: 100pF discharged through a 1.5k resistor following specification JESD22/A114.
2. Machine Model: 200pF discharged through all pins following specification JESD22/A115.

## Recommended Operating Conditions

Symbol	Rating	MIN	MAX	Unit
VDD	Supply Voltage	2.5	5.5	V
VIH	Low-level input Voltage (EN)	1.3	VDD	V
VIL	Low-level input Voltage (EN)	0	0.3	V
VIC	Common mode input voltage	0.5	VDD-0.8	V
T <sub>A</sub>	Operating free-air temperature range	-40	85	°C

## Electrical Characteristics

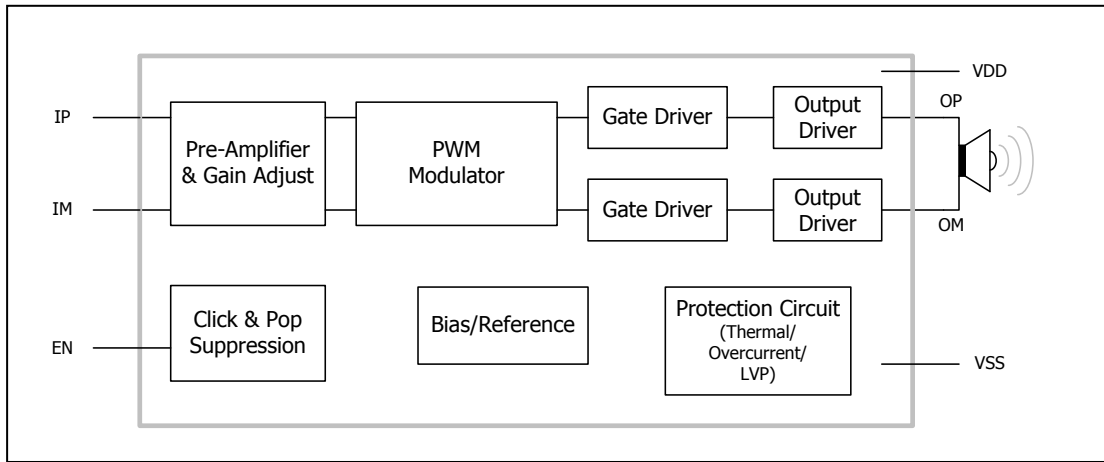
Parameter	Symbol	Conditions	MIN	TYP	MAX	Unit
Operating Supply Voltage	VDD	$T_A = -40\text{ }^\circ\text{C}$ to $+85\text{ }^\circ\text{C}$	2.5		5.5	V
Supply Quiescent Current	$I_Q$	VDD=5.5V, No load VDD=3.6V, No load VDD=2.5V, No load		3.3 2.6 2.3	5 3.5 3	mA
Shutdown Current	$I_{SD}$	VDD=2.5V to 5.5V		1	5	$\mu\text{A}$
Switching Frequency	$F_{SW}$	No input	330	380	480	kHz
Output offset Voltage	$ V_{oc} $	Inputs ac grounded, $A_V = 6\text{dB}$			25	mV
Power Supply Rejection Ratio	PSRR	VDD=2.5 to 5.5V			-75	dB
Common-mode Rejection Ratio	CMRR	Inputs shorted together VDD=2.5 to 5.5V			-60	dB

## Operating Characteristics

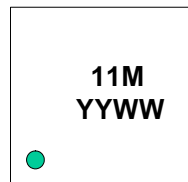
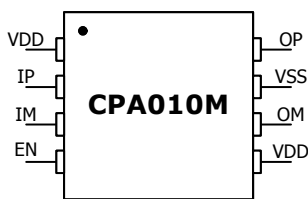
Parameter	Symbol	Conditions	TYP	MAX	Unit
Output power	$P_{OUT}$	Load=4 $\Omega$ +22 $\mu\text{H}$ , $F_{in}$ =1kHz, THD+N=1% (Note.1)	VDD=2.5	0.5	W
			VDD=3.6	1.1	
			VDD=5.5	2.6	
		Load=8 $\Omega$ +22 $\mu\text{H}$ , $F_{in}$ =1kHz, THD+N=1%	VDD=2.5	0.3	
			VDD=3.6	0.7	
			VDD=5.5	1.6	
Total harmonic distortion plus noise	THD+N	VDD=5.5V, $F_{in}$ =1kHz, Load=8 $\Omega$ , $P_{OUT}$ =0.5W	0.02	0.1	%
		VDD=3.6V, $F_{in}$ =1kHz, Load=8 $\Omega$ , $P_{OUT}$ =0.3W	0.02	0.1	
		VDD=2.5V, $F_{in}$ =1kHz, Load=8 $\Omega$ , $P_{OUT}$ =0.2W	0.02	0.1	
Signal-to-noise ratio	SNR	VDD=5V, $P_{OUT}$ =1W, Load=8 $\Omega$ , A weighted noise	96	92	dB
Output voltage noise	$V_N$	VDD=3.6V, $F_{in}$ =20Hz to 20kHz, Inputs AC grounded with $C_{IN}$ =1 $\mu\text{F}$	No weighting	50	$\mu\text{V}_{RMS}$
			A weighting	40	
Efficiency	$\eta$	Load=8 $\Omega$	VDD=5.5V	91	%
			VDD=3.6V	90	
			VDD=2.5V	89	
		Load=4 $\Omega$ (Note.1)	VDD=5.5V	86	
			VDD=3.6V	84	
			VDD=2.5V	81	
Thermal Shutdown Temperature	$T_{SD}$	Inputs ac grounded, $A_V = 6\text{dB}$	150		$^\circ\text{C}$

(Note.1) In the case of 4-Ohm load speaker, the package thermal dissipation is not enough. CPA011M is recommended to use higher impedance load speaker if the PCB artwork is not considered thermal dissipation.

## Block Diagram



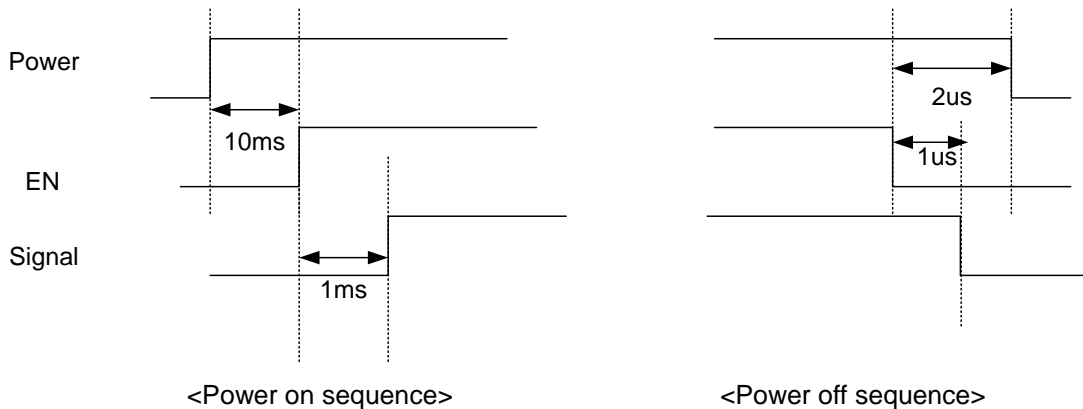
## Pin Description



11M: Product Marking  
 YYWW: Year/Work Week  
 Laser Mark

Name	I/O	PIN	Description
VDD	I	1,5	Power Supply Voltages
VSS	I	7	Power Ground Voltages
IP	I	2	Positive Audio Signal
IM	I	3	Negative Audio Signal
OP	O	8	Positive Output
OM	O	6	Negative Output
EN	I	4	Enable Control Signal(Active High)

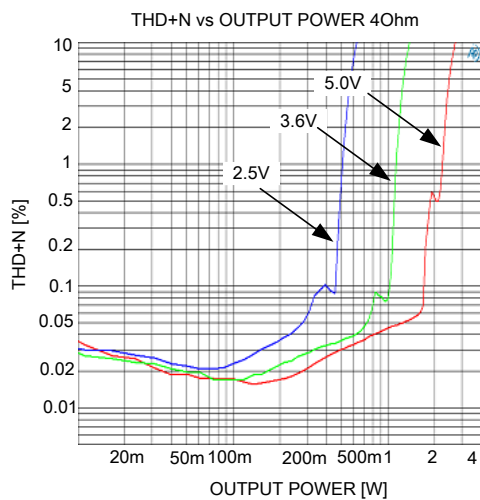
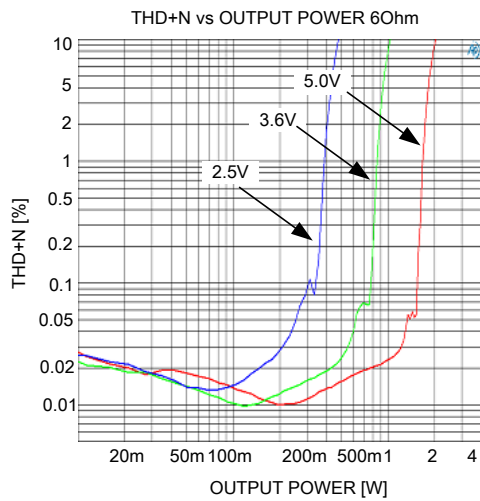
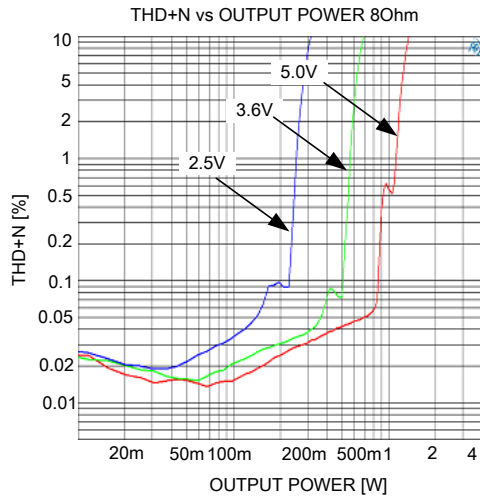
## Power On/Off Sequence



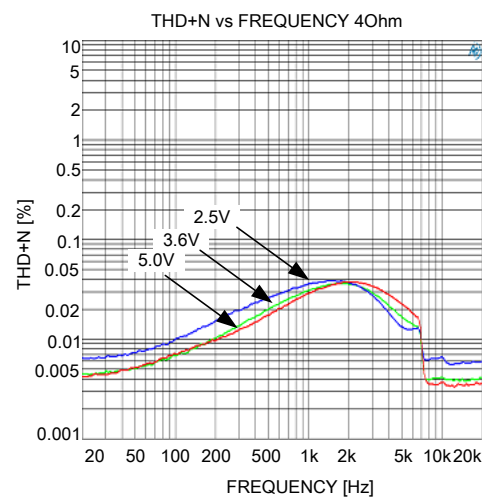
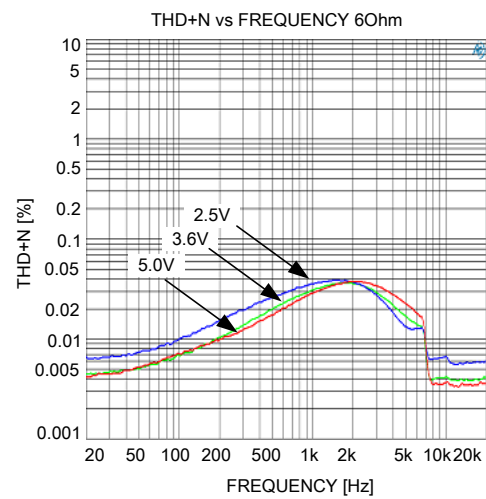
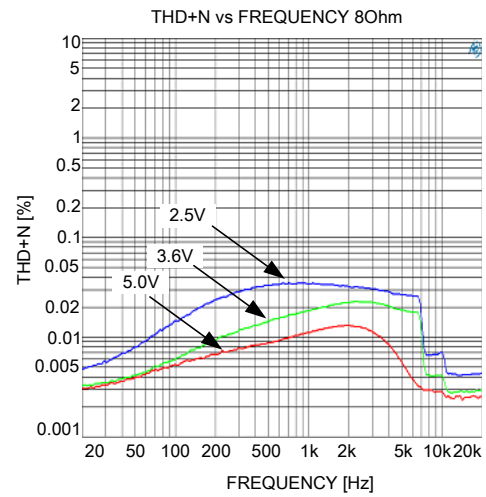
Turn on time of the audio amplifier is not longer than 1ms.

# Typical Performance Characteristics

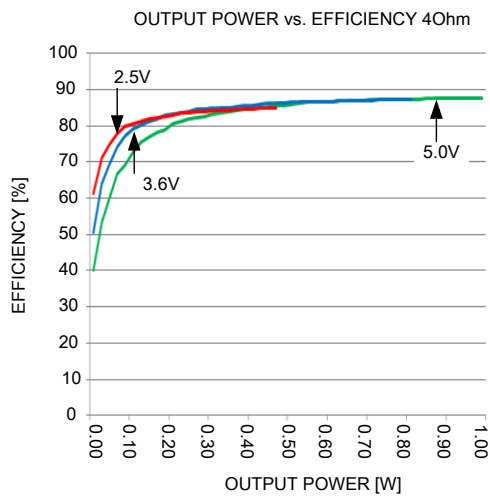
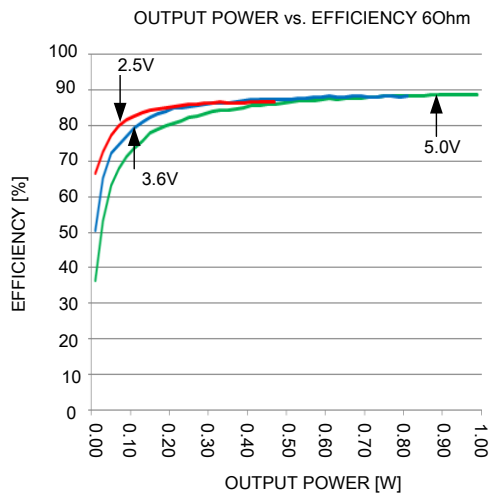
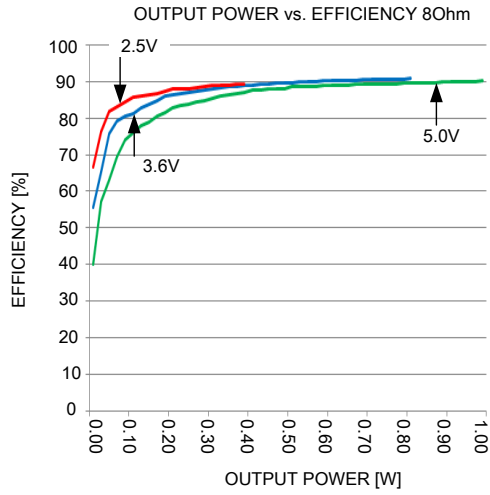
## 1. THD+N vs. Output Power



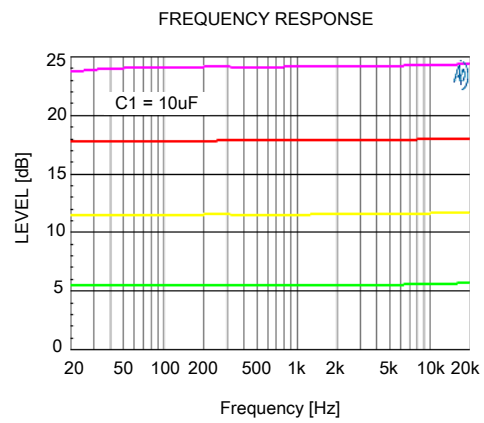
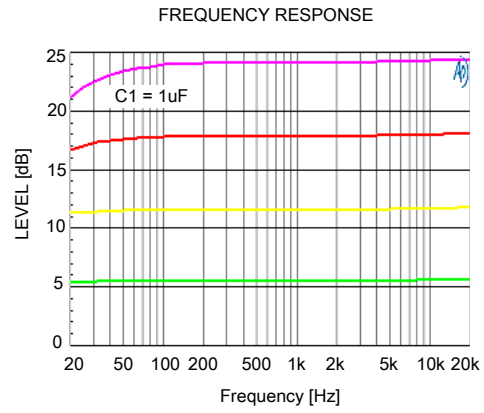
## 2. THD+N vs. Frequency



### 3. Output power vs. Efficiency



### 4. Frequency Response



## APPLICATION INFORMATION

### GENERAL AMPLIFIER FUNCTION

The CPA011M is a fully differential input/output amplifier and features a filter-less self oscillation (not using internal oscillator) and spread spectrum modulation scheme. The CPA011M requires input resistors for gain selections. The differential outputs (OP and ON) switch at about 380KHz from VDD to GND. When there is no input signal, the duty cycle of two outputs (OP and ON) is 50% in phase. Two signals cancel each other because of differential output. When there is input signal, the each pulse width of output signals (OP and ON) is changing depending on input signal amplitude. The difference of two output signals yields the differential output voltage.

### SPREAD SPECTRUM and SELF-OSCILLATION MODULATION

The CPA011M features a filter-less spread spectrum and self-oscillation modulation scheme that eliminates the need for output filter. The switching frequency varies by -40% below 400KHz frequency depending on input signal amplitude, improving EMI emissions radiated by the speaker, associated cables and traces.

The spread spectrum architecture of CPA011M spreads the energy across larger bandwidth. So switching carrier frequency does not affect the audio reproduction.

And self-oscillation scheme does not require internal oscillation, so CPA011M is effectively circuit designed.

## INPUT CIRCUIT CONFIGURATION

### Differential Input Configuration

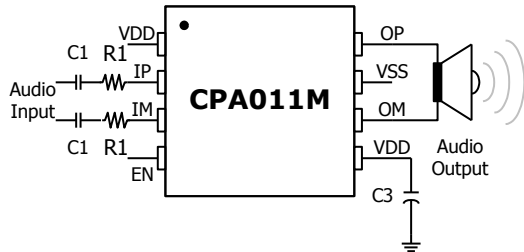


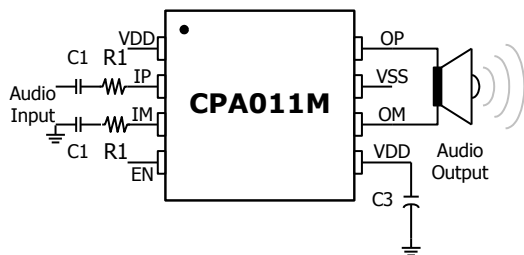
Figure2 shows the normal CPA011M differential input configuration. But if the design uses a differential source that is biased with common-mode input voltage range, input coupling capacitors are not required.

[Fig2. CPA011M Differential Input Config.]

The external resistors must be placed close to IP and IM for gain setting. Default gain  $A_v$  ( $R1 = 0 \Omega$ ) is 24dB (x16).

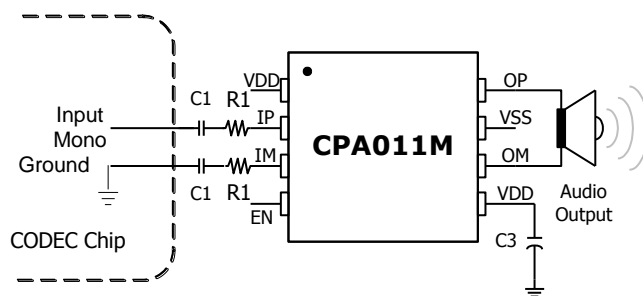
$$A_v = \frac{2 \times 80K\Omega}{(R1 + 10K\Omega)}$$

### Single Ended Input Configuration



The CPA011M can be configured as a single-ended amplifier but input capacitors are needed to block any DC at input terminal. The value of input capacitor is important to consider as it directly affects the low frequency performance.

[Fig3. CPA011M Single-Ended Input Config.#1]



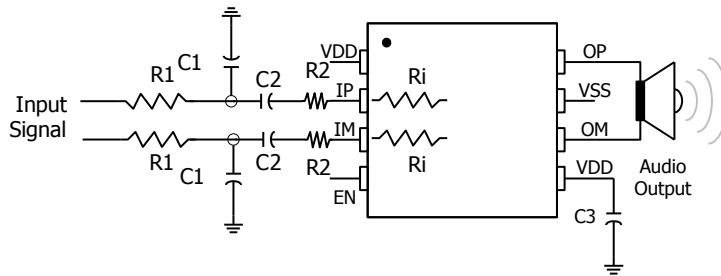
To improve audio sound quality at single-ended input configuration, IM(Negative input pin) had better connect with ground of CODEC chip. That is, noise signal of CODEC chip flows to audio amplifier, then the noise signal at CPA011M can be crossly cancelled.

[Fig4. CPA011M Single-Ended Input Config.#2]

If IM is separately grounded with CODEC chip, the noise signal of CODEC chip to IP(Positive input pin) can be amplified.



## Input Filter Design



Input filter can be sometimes be designed for reducing current consumption and improving sound quality because there are generally the constraints of overall system and the actual frequency band of interest.

**[Fig5. CPA011M Filter Design Config.]**

Although high-fidelity audio reproduction needs a flat gain response between 20Hz and 20KHz, portable devices such as cellular phone need only limited frequency audio band reproduction because of poor frequency response of speaker unit below 150Hz.

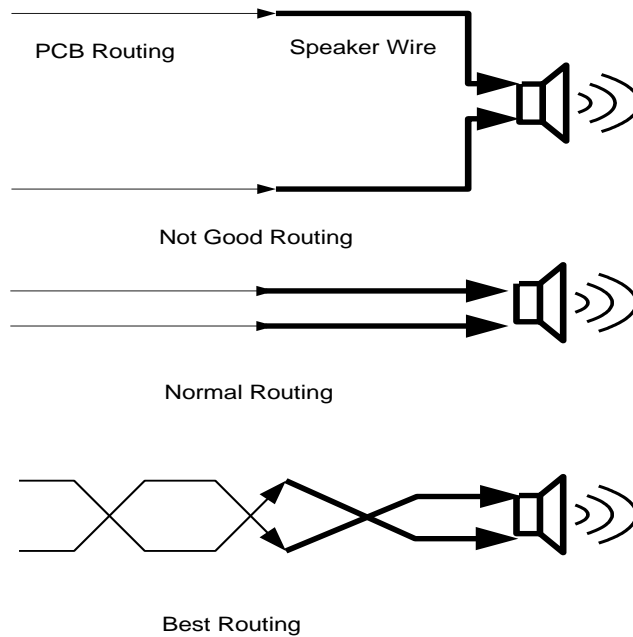
$$f_{LF} \approx \frac{1}{2\pi R1C1} \qquad f_{HF} \approx \frac{1}{2\pi(Ri + R2)C2}$$

The value of  $Ri$  is 10K $\Omega$

## OUTPUT CIRCUIT CONFIGURATION

### Output Line Placement

The CPA011M does not require an output filter (filter-less scheme) and has the high EMI immunity characteristics.

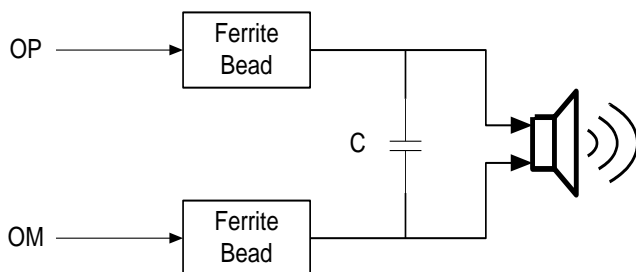


**[Fig6. Output Line Description to reduce EMI]**

To reduce the EMI, it is important that speaker line is twisted, shielded or closely paralleled.

### Output Filter

But if failing radiated emission testing without LC filter, a ferrite bead can be often used in the design. The traces from amplifier to speaker must be usually shorter because the line is functioning like RF antenna.



If choosing a ferrite bead, choose one with high impedance at high frequency, but very low impedance at low frequencies.

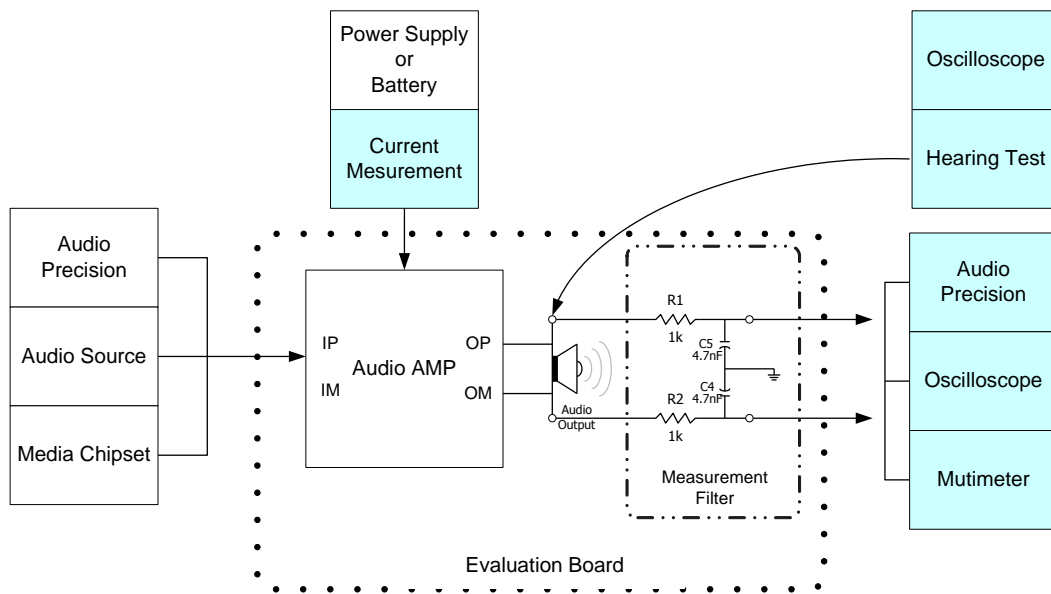
**[Fig7. Typical Ferrite Bead Filter]**

## ANOTHER IMPLEMENTATION CONSIDERATION

The external components like resistor and capacitor must be closely placed to audio amplifier. The traces between amplifier and external components distort the audio input signal and affect the power drop or fluctuation.

The use of power and ground planes will give the best THD+N performance. While reducing trace resistance, the parasitic capacitors between power and ground help to filter power supply line. The output signal line must be considered to be safely separated from another signal line. Sometimes the output line affects badly another chip's signal lines.

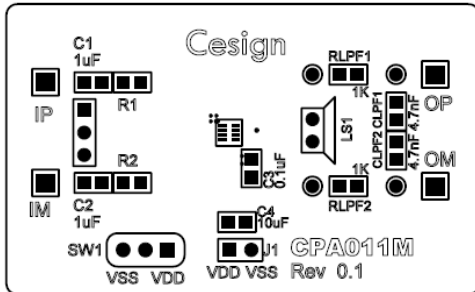
## OUTPUT MEASUREMENT ENVIRONMENT



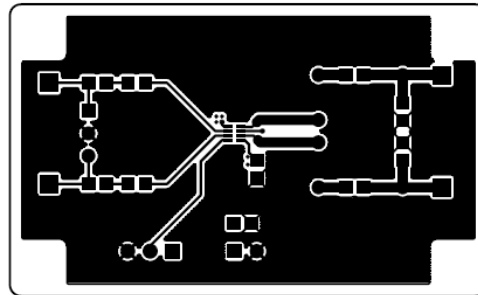
[Fig8. Test Setup Environment]

REFERENCE EVALUATION BOARD LAYOUT and SCHEMATIC

EVM Board component placement and Board Layer



< Top Placement >

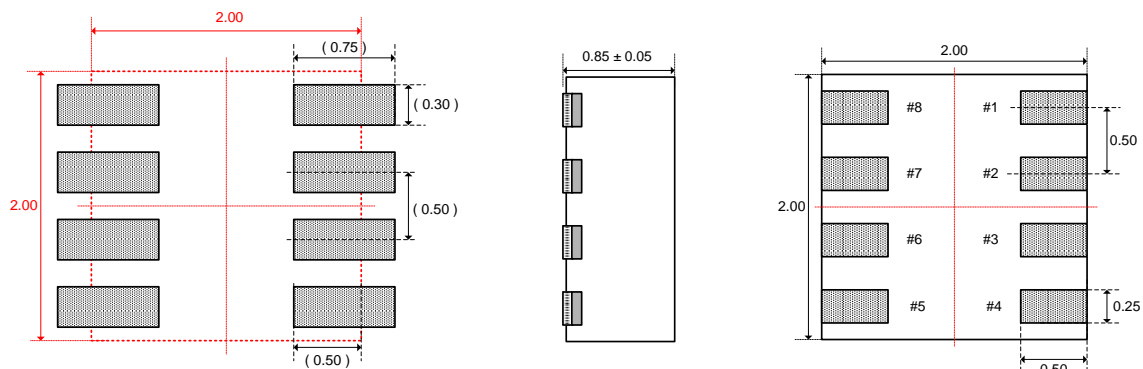


< Top Layer Routing >

Parts Descriptions

Parts	Parameter	Descriptions
C1/C2	Ceramic 1uF 0603 X7R Top place	Input DC coupling
C3	Ceramic 0.1uF 0603 X7R Top place	Power noise Reduction
C4	Ceramic 10uF 0603 X7R Top place	Power noise reduction
R1/R2	5% Chip Type	Gain Control Resistor
CLPF1/CLPF2	Ceramic 4.7nF 0603 X7R Top place	Measurement filter
RLPF1/RLPF2	Chip 1KΩ 0603 5% Top Place	Measurement filter

PHYSICAL DIMENSIONS

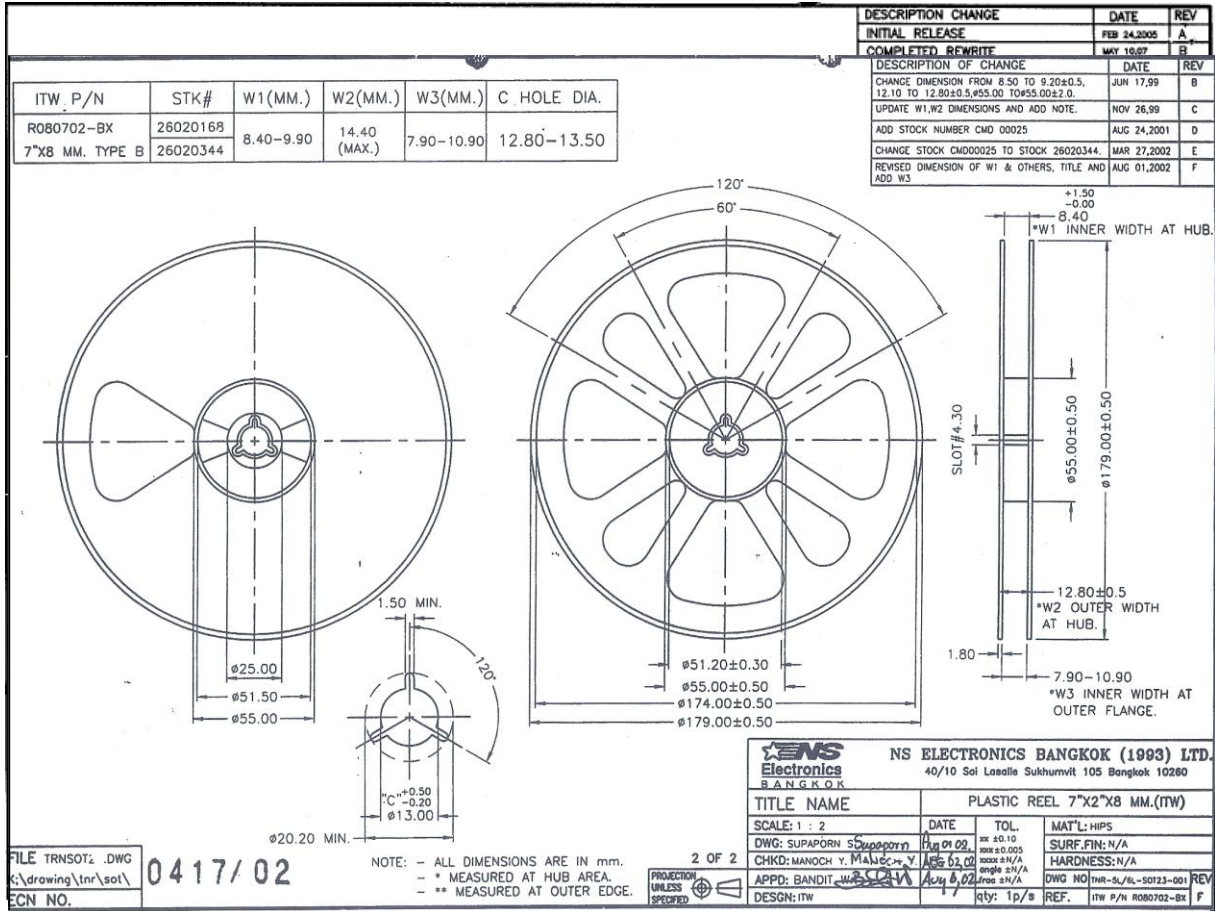


RECOMMENDED LAND PATTERN

SIDE VIEW

PACKAGE OF OUTLINE

DEMENSIONS are IN MILLIMETERS / DIMENSIONS IN ( ) FOR REFERENCE ONLY



Revision Table		
Rev No.	Date	Notes
Rev0	2007.12	The first preliminary datasheet issued
Rev0.1	2007.12.26	Application note attached
Rev0.2	2008. 2. 4	Page 1. Application Circuit Diagram changed, Page 7. External Component guide added
Rev0.3	2008. 2.26	Page13 "REFERENCE EVALUATION BOARD LAYOUT and SCHEMATIC" is added
Rev0.4	2008. 3. 12	Pin #1 VSS is changed to <b>VDD</b> for performance upgrade
Rev0.5	2008. 5. 21	EMI test Result, Making information, Physical Dimensions are added
Rev0.6	2008. 8. 7	$\overline{SD}$ → EN Pin Description changed
Rev0.7	2008. 9. 18	Power On/Off Diagram Added
Rev0.8	2010. 2. 3	THD+N, SNR, Load Condition Changed, Some items added typical value
Rev0.9	2011. 4. 6	THD+N vs Output Power, THD+N vs Frequency, Output Power vs Efficiency, and Frequency Response graph added.